#### **Digital Circuits**

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#### **Gate Circuits and Boolean Equations**

- Binary Logic and Gates

- Boolean Algebra

- Standard Forms

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## **Course Outline**

- 1. Digital Computers, Number Systems, Arithmetic Operations, Decimal, Alphanumeric, and Gray Codes
- 2. Binary Logic, Gates, Boolean Algebra, Standard Forms
- 3. Circuit Optimization, Two-Level Optimization, Map Manipulation, Multi-Level Circuit Optimization
- 4. Additional Gates and Circuits, Other Gate Types, Exclusive-OR Operator and Gates, High-Impedance Outputs
- 5. Implementation Technology and Logic Design, Design Concepts and Automation, The Design Space, Design Procedure, The major design steps
- 6. Programmable Implementation Technologies: Read-Only Memories, Programmable Logic Arrays, Programmable Array Logic, Technology mapping to programmable logic devices
- 7. Combinational Functions and Circuits
- 8. Arithmetic Functions and Circuits
- 9. Sequential Circuits Storage Elements and Sequential Circuit Analysis
- 10. Sequential Circuits, Sequential Circuit Design State Diagrams, State Tables
- 11. Counters, register cells, buses, & serial operations
- 12. Sequencing and Control, Datapath and Control, Algorithmic State Machines (ASM)
- 13. Memory Basics

# **Binary Logic and Gates**

- <u>Binary variables</u> take on one of two values.
- <u>Logical operators</u> operate on binary values and binary variables.
- Basic logical operators are the <u>logic functions</u> AND, OR and NOT.
- <u>Logic gates</u> implement logic functions.
- <u>Boolean Algebra</u>: a useful mathematical system for specifying and transforming logic functions.
- We study Boolean algebra as foundation for designing and analyzing digital systems!

# **Binary Variables**

- Recall that the two binary values have different names:
  - True/False
  - On/Off
  - Yes/No
  - 1/0
- We use 1 and 0 to denote the two values.
- Variable identifier examples:
  - $-A, B, y, z, or X_1$  for now
  - RESET, START\_IT, or ADD1 later

# **Logical Operations**

- The three basic logical operations are:
  - AND
  - -OR
  - NOT
- AND is denoted by a dot  $(\cdot)$
- OR is denoted by a plus (+)
- NOT is denoted by an overbar (<sup>-</sup>), a single quote mark (') after, or (~) before the variable

## **Notation Examples**

- Examples:
- Y=A.B is read "Y is equal to A AND B."
- z=x+y is read "z is equal to x OR y."
- $X=\overline{A}$  is read "X is equal to NOT A."
- Note:
  - The statement:
    - 1 + 1 = 2 (read "one plus one equals two")
      is not the same as
    - 1 + 1 = 1 (read "1 or 1 equals 1").

## **Operator Definitions**

• Operations are defined on the values "0" and "1" for each operator:

ANDORNOT $0 \cdot 0 = 0$ 0 + 0 = 0 $\overline{0} = 1$  $0 \cdot 1 = 0$ 0 + 1 = 1 $\overline{1} = 0$  $1 \cdot 0 = 0$ 1 + 0 = 1 $1 \cdot 1 = 1$ 1 + 1 = 1

## **Truth Tables**

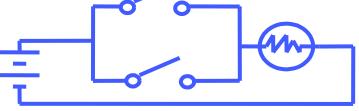
- Truth table
  - a tabular listing of the values of a function for all possible combinations of values on its arguments
- Example:
  - Truth tables for the basic logic operations:

AND				OR			NOT	
X	Y	$\mathbf{Z} = \mathbf{X} \cdot \mathbf{Y}$	X	Y	$\mathbf{Z} = \mathbf{X} + \mathbf{Y}$		X	$Z = \overline{X}$
0	0	0	0	0	0		0	1
0	1	0	0	1	1		1	0
1	0	0	1	0	1			
1	1	1	1	1	1			

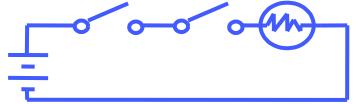
# **Logic Function Implementation**

- Using Switches
  - For inputs:
    - logic 1 is switch closed
    - logic 0 is <u>switch open</u>
  - For outputs:
    - logic 1 is <u>light on</u>
    - logic 0 is <u>light off</u>.
  - NOT uses a switch such that:
    - logic 1 is switch open
    - logic 0 is <u>switch closed</u>

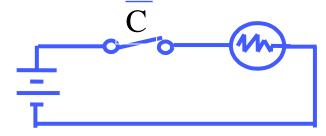




Switches in series => AND

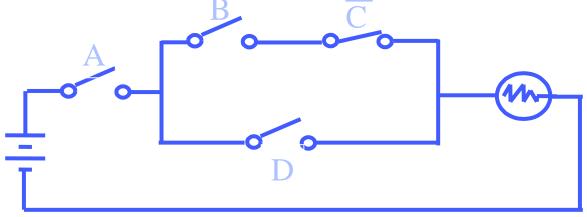


Normally-closed switch => NOT



#### Logic Function Implementation (Continued)

• Example: Logic Using Switches



• Light is on (L = 1) for

 $L(A, B, C, D) = A \cdot ((B \cdot C') + D)$ and off (L = 0), otherwise.

• Useful model for relay circuits and for CMOS gate circuits, the foundation of current digital logic technology

# **Logic Gates**

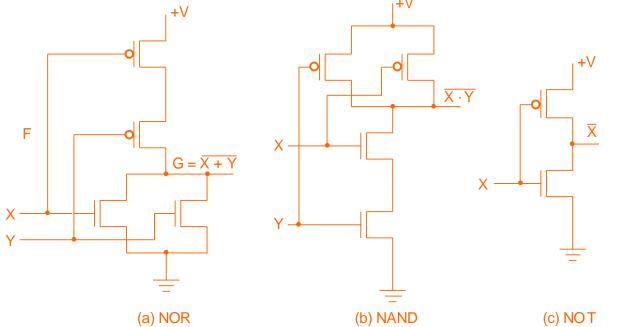
• In the earliest computers, switches were opened and closed by magnetic fields produced by energizing coils in *relays*.

The switches in turn opened and closed the current paths.

- Later, *vacuum tubes* that open and close current paths electronically replaced relays.
- Today, *transistors* are used as electronic switches that open and close current paths.

# Logic Gates (continued)

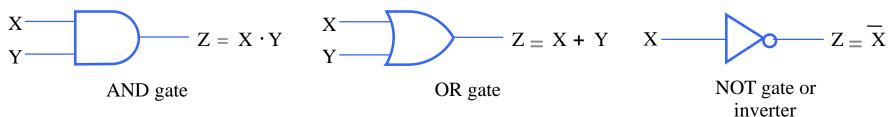
 Implementation of logic gates with transistors (See Reading Supplement – CMOS Circuits)



- Transistor or tube implementations of logic functions are called logic gates or just gates
- Transistor gate circuits can be modeled by switch circuits

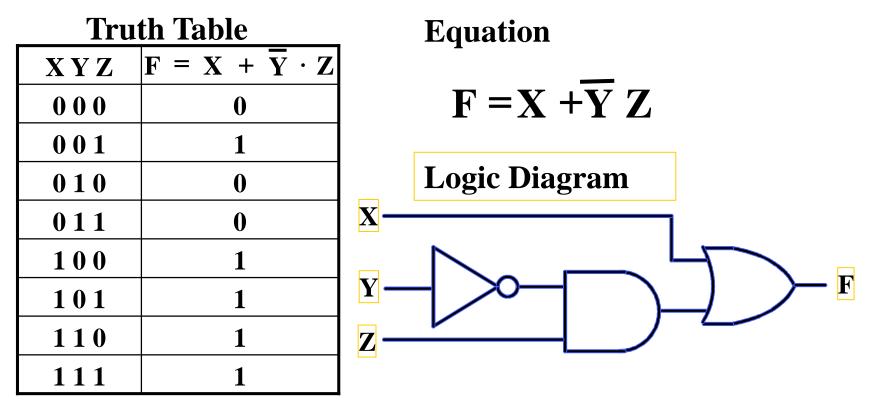
# **Logic Gate Symbols and Behavior**

• Logic gates have special symbols (Graphic symbols ):



• Waveform behavior in time (Timing diagram ):

# **Logic Diagrams and Expressions**



- Boolean equations, truth tables and logic diagrams describe the same function!
- Truth tables are unique; expressions and logic diagrams are not. This gives flexibility in implementing functions.

## **Boolean Algebra**

An algebraic structure defined on a set of at least two elements, B, together with three binary operators (denoted +, · and ) that satisfies the following basic identities:

1.	X + 0 = X	2.	$X \cdot 1 = X$	<b>Existence of 0 and 1</b>
3.	X + 1 = 1	4.	$X \cdot 0 = 0$	
5.	X + X = X	6.	$X \cdot X = X$	Idempotence
7.	$X + \overline{X} = 1$	8.	$X \cdot \overline{X} = 0$	<b>Existence of complement</b>
9.	$\overline{\overline{X}} = X$			Involution
10.	X + Y = Y + X	11.	XY = YX	Commutative
12.	(X+Y)+Z = X+(Y+Z)	13.	(XY)Z = X(	YZ) Associative
14.	X(Y+Z) = XY+XZ	15.	X + YZ = (2	(X + Y)(X + Z) Distributive
16.	$\overline{X+Y} = \overline{X} \cdot \overline{Y}$	17.	$\overline{X \cdot Y} = \overline{X} + \overline{X}$	Y DeMorgan's

#### Some Properties of Identities & the Algebra

- If the meaning is unambiguous, we leave out the symbol "·"
- The identities appear in **dual** pairs.
- The dual of an algebraic expression is obtained by interchanging + and • and interchanging 0's and 1's.
  - A function is said to be self-dual if and only if its dual is equivalent to the given function,
    - i.e., if a given function is f(X, Y, Z) = (XY + YZ + ZX) then its dual is, fd(X, Y, Z) = (X + Y).(Y + Z).(Z + X) = (XY + YZ + ZX), it is equivalent to the given function

#### Some Properties of Identities & the Algebra

• Unless it happens to be self-dual, the dual of an expression does not equal the expression itself.

• Example: 
$$F = (A + \overline{C}) \cdot B + 0$$
  
dual  $F = (A \cdot \overline{C} + B) \cdot 1 = A \cdot \overline{C} + B$ 

- Example:  $G = X \cdot Y + (\overline{W + Z})$ dual G =
- Example:  $H = A \cdot B + A \cdot C + B \cdot C$ dual H =
- Are any of these functions self-dual?

#### Some Properties of Identities & the Algebra

- There can be more than 2 elements in B,
  - i. e., elements other than 1 and 0.
- What are some common useful Boolean algebras with more than 2 elements?
  - Algebra of Sets
  - Algebra of n-bit binary vectors
- If B contains only 1 and 0, then B is called the <u>switching algebra</u> which is the algebra we use most often.

# **Boolean Operator Precedence**

- The order of evaluation in a Boolean expression is:
  - Parentheses
  - NOT
  - AND
  - OR
- Consequence:
  - Parentheses appear around OR expressions
- Example:
  - $F = A(B + C)(C + \overline{D})$

# **Example 1: Boolean Algebraic Proof**

- $A + A \cdot B = A$  (Absorption Theorem) Proof Steps Justification (identity or theorem)
- $A + A \cdot B$ =  $A \cdot 1 + A \cdot B$   $X = X \cdot 1$ =  $A \cdot (1 + B)$   $X \cdot Y + X \cdot Z = X \cdot (Y + Z)$  (Distributive Law) =  $A \cdot 1$  1 + X = 1= A  $X \cdot 1 = X$
- Our primary reason for doing proofs is to learn:
  - Careful and efficient use of the identities and theorems of Boolean algebra, and
  - How to choose the appropriate identity or theorem to apply to make forward progress, irrespective of the application.

# **Example 2: Boolean Algebraic Proofs**

- $AB + \overline{A}C + BC = AB + \overline{A}C$  (Consensus Theorem)
- Proof Steps: Justification (identity or theorem)  $AB + \overline{A}C + BC$ 
  - $= AB + \overline{A}C + 1 \cdot BC$
  - $= AB + \overline{A}C + (A + \overline{A}) \cdot BC$
  - $= AB + \overline{AC} + ABC + \overline{ABC}$
  - $= AB(1 + C) + \overline{A}C(1 + B)$
  - $= AB \cdot 1 + \overline{A}C \cdot 1$
  - $= AB + \overline{A}C$

## **Example 3: Boolean Algebraic Proofs**

#### • $(\overline{X + Y})Z + X\overline{Y} = \overline{Y}(X + Z)$ Proof Steps Justification (identity or theorem) $(\overline{X + Y})Z + X\overline{Y}$

=

#### **Useful Theorems**

 $x \cdot y + \overline{x} \cdot y = y \quad (x + y)(\overline{x} + y) = y \quad \text{Minimization}$   $x + x \cdot y = x \quad x \cdot (x + y) = x \quad \text{Absorption}$   $x + \overline{x} \cdot y = x + y \quad x \cdot (\overline{x} + y) = x \cdot y \quad \text{Simplification}$   $x \cdot y + \overline{x} \cdot z + y \cdot z = x \cdot y + \overline{x} \cdot z \quad \text{Consensus}$   $(x + y) \cdot (\overline{x} + z) \cdot (y + z) = (x + y) \cdot (\overline{x} + z)$   $\overline{x + y} = \overline{x} \cdot \overline{y} \quad \overline{x \cdot y} = \overline{x} + \overline{y} \quad \text{DeMorgan's Laws}$ 

#### **Proof of Simplification**

# $\mathbf{x} \cdot \mathbf{y} + \overline{\mathbf{x}} \cdot \mathbf{y} = \mathbf{y}$

# $(x+y)(\overline{x}+y) = y$

#### **Proof of DeMorgan's Laws**

# $\overline{\mathbf{x} + \mathbf{y}} = \overline{\mathbf{x}} \cdot \overline{\mathbf{y}}$

# $\overline{\mathbf{x}\cdot\mathbf{y}} = \overline{\mathbf{x}} + \overline{\mathbf{y}}$

#### **Boolean Function Evaluation**

F1 =  $xy\overline{z}$ F2 =  $x + \overline{y}z$ F3 =  $\overline{x}\overline{y}\overline{z} + \overline{x}yz + x\overline{y}$ F4 =  $x\overline{y} + \overline{x}z$ 

X	y	Z	<b>F1</b>	<b>F2</b>	<b>F3</b>	<b>F4</b>
0	0	0	0	0	1	0
0	0	1	0	1	0	1
0	1	0	0	0	0	0
0	1	1	0	0	1	1
1	0	0	0	1	1	1
1	0	1	0	1	1	1
1	1	0	1	1	0	0
1	1	1	0	1	0	0

# **Expression Simplification**

- An application of Boolean algebra
- Simplify to contain the smallest number of <u>literals</u> (complemented and uncomplemented variables):
  - $\mathbf{A}\mathbf{B} + \overline{\mathbf{A}}\mathbf{C}\mathbf{D} + \overline{\mathbf{A}}\mathbf{B}\mathbf{D} + \overline{\mathbf{A}}\mathbf{C}\overline{\mathbf{D}} + \mathbf{A}\mathbf{B}\mathbf{C}\mathbf{D}$
- $= \mathbf{A}\mathbf{B} + \mathbf{A}\mathbf{B}\mathbf{C}\mathbf{D} + \mathbf{\overline{A}}\mathbf{C}\mathbf{D} + \mathbf{\overline{A}}\mathbf{C}\mathbf{\overline{D}} + \mathbf{\overline{A}}\mathbf{B}\mathbf{D}$
- $= \mathbf{A}\mathbf{B} + \mathbf{A}\mathbf{B}(\mathbf{C}\mathbf{D}) + \mathbf{\overline{A}}\mathbf{C}(\mathbf{D} + \mathbf{\overline{D}}) + \mathbf{\overline{A}}\mathbf{B}\mathbf{D}$
- $= \mathbf{A}\mathbf{B} + \overline{\mathbf{A}}\mathbf{C} + \overline{\mathbf{A}}\mathbf{B}\mathbf{D} = \mathbf{B}(\mathbf{A} + \overline{\mathbf{A}}\mathbf{D}) + \overline{\mathbf{A}}\mathbf{C}$
- $= B (A + D) + \overline{A}C$  5 literals

# **Complementing Functions**

- Use DeMorgan's Theorem to complement a function:
  - **1. Interchange AND and OR operators**
  - 2. Complement each constant value and literal
- Example: Complement F =  $\overline{x}y\overline{z}+x\overline{y}\overline{z}$  $\overline{F} = (x + \overline{y} + z)(\overline{x} + y + z)$
- Example: Complement  $G = (\overline{a} + bc)\overline{d} + e$  $\overline{G} = ?$

## **Overview – Canonical Forms**

- What are Canonical Forms?
- Minterms and Maxterms
- Index Representation of Minterms and Maxterms
- Sum-of-Minterm (SOM) Representations
- Product-of-Maxterm (POM) Representations
- Representation of Complements of Functions
- Conversions between Representations

## **Canonical Forms**

- It is useful to specify Boolean functions in a form that:
  - Allows comparison for equality.
  - Has a correspondence to the truth tables
- Canonical Forms in common usage:
  - Sum of Minterms (SOM)
  - Product of Maxterms (POM)

## Minterms

- <u>Minterms</u> are AND terms with every variable present in either true or complemented form.
- Given that each binary variable may appear normal (e.g., x) or complemented (e.g., x), there are 2<sup>n</sup> minterms for *n* variables.
- <u>Example</u>: Two variables (X and Y)produce  $2 \ge 2 = 4$  combinations:
  - **XY** (both normal)
  - **XY** (X normal, Y complemented)
  - **XY** (X complemented, Y normal)
  - **XY** (both complemented)
- Thus there are <u>four minterms</u> of two variables.

#### Maxterms

- <u>Maxterms</u> are OR terms with every variable in true or complemented form.
- Given that each binary variable may appear normal (e.g., x) or complemented (e.g., x), there are 2<sup>n</sup> maxterms for *n* variables.
- <u>Example:</u> Two variables (X and Y) produce 2 x 2 = 4 combinations:

X+Y	(both normal)
$X+\overline{Y}$	(X normal, Y complemented)
<b>X</b> +Y	(X complemented, Y normal)
$\overline{\mathbf{X}} + \overline{\mathbf{Y}}$	(both complemented)

#### **Maxterms and Minterms**

• Examples: Two variable minterms and maxterms.

Index	Minterm	Maxterm
0	$\overline{\mathbf{x}} \overline{\mathbf{y}}$	<b>x</b> + <b>y</b>
1	x y	$\mathbf{x} + \overline{\mathbf{y}}$
2	xy	$\overline{\mathbf{x}} + \mathbf{y}$
3	x y	$\overline{\mathbf{x}} + \overline{\mathbf{y}}$

• The index above is important for describing which variables in the terms are true and which are complemented.

## **Standard Order**

- Minterms and maxterms are designated with a subscript
- The subscript is a number, corresponding to a binary pattern
- The bits in the pattern represent the complemented or normal state of each variable listed in a standard order.
- All variables will be present in a minterm or maxterm and will be listed in the <u>same order</u> (usually alphabetically)
- Example: For variables a, b, c:
  - Maxterms:  $(a + b + \bar{c})$ , (a + b + c)
  - Terms: (b + a + c), a  $\bar{c}$  b, and (c + b + a) are NOT in standard order.
  - Minterms:  $a \bar{b} c$ , a b c,  $\bar{a} \bar{b} c$
  - Terms: (a + c), b c, and (a + b) do not contain all variables

## **Purpose of the Index**

- The <u>index</u> for the minterm or maxterm, expressed as a binary number, is used to determine whether the variable is shown in the true form or complemented form.
- For Minterms:
  - "1" means the variable is "Not Complemented" and
  - "0" means the variable is "Complemented".
- For Maxterms:
  - "0" means the variable is "Not Complemented" and
  - "1" means the variable is "Complemented".

## **Index Example in Three Variables**

- Example: (for three variables)
- Assume the variables are called X, Y, and Z.
- The standard order is X, then Y, then Z.
- The <u>Index 0</u> (base 10) = 000 (base 2) for three variables). All three variables are complemented for <u>minterm 0</u> ( $\overline{X}, \overline{Y}, \overline{Z}$ ) and no variables are complemented for <u>Maxterm 0</u> (X,Y,Z).
  - Minterm 0, called  $m_0$  is  $\overline{X}\overline{Y}\overline{Z}$ .
  - Maxterm 0, called  $M_0$  is (X + Y + Z).
  - Minterm 6 ?
  - Maxterm 6 ?

#### **Index Examples – Four Variables**

#### **Index Binary Minterm Maxterm**

i	Pattern	$\mathbf{m}_{\mathbf{i}}$	$\mathbf{M_{i}}$
0	0000	abcd	a+b+c+d
1	0001	abcd	?
3	0011	?	$a+b+\overline{c}+\overline{d}$
5	0101	abcd	$\mathbf{a} + \mathbf{\overline{b}} + \mathbf{c} + \mathbf{\overline{d}}$
7	0111	?	$\mathbf{a} + \mathbf{\overline{b}} + \mathbf{\overline{c}} + \mathbf{\overline{d}}$
10	1010	abcd	$\bar{a}+b+\bar{c}+d$
13	1101	abīd	?
15	1111	abcd	$\overline{a} + \overline{b} + \overline{c} + \overline{d}$

# **Minterm and Maxterm Relationship**

- Review: DeMorgan's Theorem  $\overline{x \cdot y} = \overline{x} + \overline{y}$  and  $\overline{x + y} = \overline{x} \cdot \overline{y}$
- Two-variable example:  $M_2 = \overline{x} + y$  and  $m_2 = x \cdot \overline{y}$ Thus  $M_2$  is the complement of  $m_2$  and vice-versa.
- Since DeMorgan's Theorem holds for *n* variables, the above holds for terms of *n* variables
- giving:

$$\mathbf{M}_{i} = \overline{\mathbf{m}}_{i \text{ and }} \mathbf{m}_{i} = \overline{\mathbf{M}}_{i}$$

Thus M<sub>i</sub> is the complement of m<sub>i</sub>.

### **Function Tables for Both**

Minterms of Maxterms of 2 variables
2 variables

x y	m <sub>0</sub>	<b>m</b> <sub>1</sub>	<b>m</b> <sub>2</sub>	<b>m</b> <sub>3</sub>	ſ	ху	$\mathbf{M}_{0}$	$\mathbf{M}_{1}$	<b>M</b> <sub>2</sub>	M3
00	1	0	0	0		00	0	1	1	1
01	0	1	0	0	ſ	01	1	0	1	1
10	0	0	1	0		10	1	1	0	1
11	0	0	0	1		11	1	1	1	0

 Each column in the maxterm function table is the complement of the column in the minterm function table since M<sub>i</sub> is the complement of m<sub>i</sub>.

## **Observations**

- In the function tables:
  - Each <u>min</u>term has one and only one 1 present in the  $2^n$  terms (a <u>minimum</u> of 1s). All other entries are 0.
  - Each <u>max</u>term has one and only one 0 present in the  $2^n$  terms All other entries are 1 (a <u>max</u>imum of 1s).
- We can implement any function by "ORing" the minterms corresponding to "1" entries in the function table. These are called the minterms of the function.
- We can implement any function by "ANDing" the maxterms corresponding to "0" entries in the function table. These are called the maxterms of the function.
- This gives us two <u>canonical forms</u>:
  - <u>Sum of Minterms (SOM)</u>
  - Product of Maxterms (POM)

for stating any Boolean function.

#### **Minterm Function Example**

- Example: Find  $F_1 = m_1 + m_4 + m_7$
- $F1 = \overline{x} \overline{y} z + x \overline{y} \overline{z} + x y z$

хуz	index	<b>m</b> <sub>1</sub>	+	m <sub>4</sub>	+	<b>m</b> <sub>7</sub>	$= \mathbf{F}_1$
000	0	0	+	0	+	0	= 0
001	1	1	+	0	+	0	= 1
010	2	0	+	0	+	0	= 0
011	3	0	+	0	+	0	= 0
100	4	0	+	1	+	0	= 1
101	5	0	+	0	+	0	= 0
110	6	0	+	0	+	0	= 0
111	7	0	+	0	+	1	= 1

#### **Minterm Function Example**

- $F(A, B, C, D, E) = m_2 + m_9 + m_{17} + m_{23}$
- F(A, B, C, D, E) =

#### **Maxterm Function Example**

• Example: Implement F1 in maxterms:  $F_1 = M_0 \cdot M_2 \cdot M_3 \cdot M_5 \cdot M_6$   $F_1 = (x + y + z) \cdot (x + \overline{y} + z) \cdot (x + \overline{y} + \overline{z})$  $\cdot (\overline{x} + y + \overline{z}) \cdot (\overline{x} + \overline{y} + z)$ 

хуz	i	$\mathbf{M}_0 \cdot \mathbf{M}_2 \cdot \mathbf{M}_3 \cdot \mathbf{M}_5 \cdot \mathbf{M}_6 = \mathbf{F1}$
000	0	$0 \cdot 1 \cdot 1 \cdot 1 \cdot 1 = 0$
001	1	$1 \cdot 1 \cdot 1 \cdot 1 \cdot 1 = 1$
010	2	$1 \cdot 0 \cdot 1 \cdot 1 \cdot 1 = 0$
011	3	$1 \cdot 1 \cdot 0 \cdot 1 \cdot 1 = 0$
100	4	$1 \cdot 1 \cdot 1 \cdot 1 \cdot 1 = 1$
101	5	$1 \cdot 1 \cdot 1 \cdot 0 \cdot 1 = 0$
110	6	$1 \cdot 1 \cdot 1 \cdot 1 \cdot 0 = 0$
111	7	$1 \cdot 1 \cdot 1 \cdot 1 \cdot 1 = 1$

#### **Maxterm Function Example**

- $\mathbf{F}(\mathbf{A},\mathbf{B},\mathbf{C},\mathbf{D}) = \mathbf{M}_3 \cdot \mathbf{M}_8 \cdot \mathbf{M}_{11} \cdot \mathbf{M}_{14}$
- F(A, B, C, D) =

## **Canonical Sum of Minterms**

- Any Boolean function can be expressed as a <u>Sum of Minterms</u>.
  - For the function table, the <u>minterms</u> used are the terms corresponding to the 1's
  - For expressions, <u>expand</u> all terms first to explicitly list all minterms. Do this by "ANDing" any term missing a variable v with a term  $(v + \overline{v})$ .
- Example: Implement  $f = x + \overline{x} \overline{y}$  as a sum of minterms.

First expand terms:  $\mathbf{f} = \mathbf{x}(\mathbf{y} + \overline{\mathbf{y}}) + \overline{\mathbf{x}} \ \overline{\mathbf{y}}$ Then distribute terms:  $\mathbf{f} = \mathbf{x}\mathbf{y} + \mathbf{x}\overline{\mathbf{y}} + \overline{\mathbf{x}}\ \overline{\mathbf{y}}$ Express as sum of minterms:  $\mathbf{f} = \mathbf{m}_3 + \mathbf{m}_2 + \mathbf{m}_0$ 

# **Another SOM Example**

- Example:  $\mathbf{F} = \mathbf{A} + \overline{\mathbf{B}} \mathbf{C}$
- There are three variables, A, B, and C which we take to be the standard order.
- Expanding the terms with missing variables:

- Collect terms (removing all but one of duplicate terms):
- Express as SOM:

## **Shorthand SOM Form**

- From the previous example, we started with:  $F = A + \overline{B}C$
- We ended up with:
- $F = m_1 + m_4 + m_5 + m_6 + m_7$
- This can be denoted in the formal shorthand:  $F(A,B,C) = \Sigma_m(1,4,5,6,7)$
- Note that we explicitly show the standard variables in order and drop the "m" designators.

## **Canonical Product of Maxterms**

- Any Boolean Function can be expressed as a <u>Product</u> of <u>Maxterms (POM)</u>.
  - For the function table, the maxterms used are the terms corresponding to the 0's.
  - For an expression, expand all terms first to explicitly list all maxterms. Do this by first applying the second distributive law, "ORing" terms missing variable V with a term equal to V·V and then applying the distributive law again.
- Example: Convert to product of maxterms:

$$f(x, y, z) = x + \overline{x} \overline{y}$$
  
Apply the distributive law:

 $x + \overline{x} \overline{y} = (x + \overline{x})(x + \overline{y}) = 1 \cdot (x + \overline{y}) = x + \overline{y}$ 

Add missing variable z:

$$\mathbf{x} + \overline{\mathbf{y}} + \mathbf{z} \cdot \overline{\mathbf{z}} = (\mathbf{x} + \overline{\mathbf{y}} + \mathbf{z})(\mathbf{x} + \overline{\mathbf{y}} + \overline{\mathbf{z}})$$

**Express as POM:**  $f = M_2 \cdot M_3$ 

## **Another POM Example**

• Convert to Product of Maxterms:

 $f(A,B,C) = A \overline{C} + BC + \overline{A} \overline{B}$ 

- Use  $x + y z = (x+y) \cdot (x+z)$  with x = (AC + BC), y = A, and  $z = \overline{B}$  to get:  $f = (A\overline{C} + BC + \overline{A})(A\overline{C} + BC + \overline{B})$
- Then use  $x + \overline{x}y = x + y$  to get:  $f = (\overline{C} + BC + \overline{A})(A\overline{C} + C + \overline{B})$

and a second time to get:

$$\mathbf{f} = (\overline{\mathbf{C}} + \mathbf{B} + \overline{\mathbf{A}})(\mathbf{A} + \mathbf{C} + \overline{\mathbf{B}})$$

• Rearrange to standard order,

 $f = (\overline{A} + B + \overline{C})(A + \overline{B} + C)$  to give  $f = M_5 \cdot M_2$ 

## **Function Complements**

- The complement of a function expressed as a sum of minterms is constructed by selecting the minterms missing in the sum-of-minterms canonical forms.
- Alternatively, the complement of a function expressed by a Sum of Minterms form is simply the Product of Maxterms with the same indices.
- Example: Given  $F(x, y, z) = \Sigma_m(1, 3, 5, 7)$  $\overline{F}(x, y, z) = \Sigma_m(0, 2, 4, 6)$  $\overline{F}(x, y, z) = \Pi_M(1, 3, 5, 7)$

## **Conversion Between Forms**

- To convert between sum-of-minterms and productof-maxterms form (or vice-versa) we follow these steps:
  - Find the function complement by swapping terms in the list with terms not in the list.
  - Change from products to sums, or vice versa.
- Example: Given F as before:  $F(x, y, z) = \Sigma_m(1, 3, 5, 7)$
- Form the Complement:  $\overline{F}(x, y, z) = \Sigma_m(0, 2, 4, 6)$
- Then use the other form with the same indices this forms the complement again, giving the other form of the original function:  $F(x, y, z) = \prod_M (0, 2, 4, 6)$

## **Standard Forms**

- <u>Standard Sum-of-Products (SOP) form:</u> equations are written as an OR of AND terms
- <u>Standard Product-of-Sums (POS) form:</u> equations are written as an AND of OR terms
- Examples:
  - **SOP:**  $A B C + \overline{A} \overline{B} C + B$
  - **POS:**  $(\mathbf{A}+\mathbf{B})\cdot(\mathbf{A}+\mathbf{\overline{B}}+\mathbf{\overline{C}})\cdot\mathbf{C}$
- These "mixed" forms are <u>neither SOP nor POS</u>
  - (A B + C) (A + C)
  - $AB\overline{C} + AC(A+B)$

# **Standard Sum-of-Products (SOP)**

- A sum of minterms form for *n* variables can be written down directly from a truth table.
  - Implementation of this form is a two-level network of gates such that:
  - The first level consists of *n*-input AND gates, and
  - The second level is a single OR gate (with fewer than  $2^n$  inputs).
- This form often can be simplified so that the corresponding circuit is simpler.

## **Standard Sum-of-Products (SOP)**

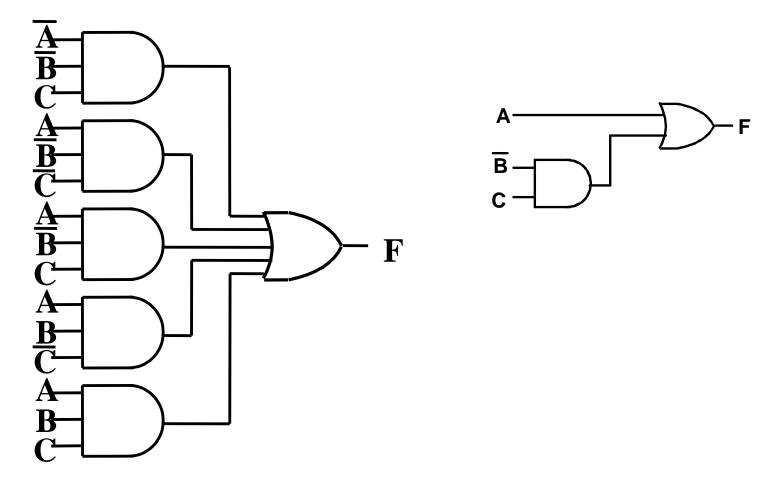
- A Simplification Example:
- $F(A,B,C) = \Sigma m(1,4,5,6,7)$
- Writing the minterm expression:  $F = \overline{A} \overline{B} C + A \overline{B} \overline{C} + A \overline{B} \overline{C} + AB\overline{C} + AB\overline{C} + AB\overline{C}$
- Simplifying:

 $\mathbf{F} = \mathbf{A} + \overline{\mathbf{B}}\mathbf{C}$ 

 Simplified F contains 3 literals compared to 15 in minterm F

#### **AND/OR Two-level Implementation of SOP Expression**

• The two implementations for F are shown below – it is quite apparent which is simpler!



- The previous examples show that:
  - Canonical Forms (Sum-of-minterms, Product-of-Maxterms), or other standard forms (SOP, POS) differ in complexity
  - Boolean algebra can be used to manipulate equations into simpler forms.
  - Simpler equations lead to simpler two-level implementations

#### • Questions:

- How can we attain a "simplest" expression?
- Is there only one minimum cost circuit?
- The next part will deal with these issues.