

## Digital Signal Processing

Prof. Nizamettin AYDIN

[naydin@yildiz.edu.tr](mailto:naydin@yildiz.edu.tr)

<http://www.yildiz.edu.tr/~naydin>

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## Digital Signal Processing

Lecture 15

### Zeros of $H(z)$ and the Frequency Domain

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### READING ASSIGNMENTS

- This Lecture:
  - Chapter 7, Section 7-6 to end
- Other Reading:
  - Recitation & Lab: Chapter 7
    - ZEROS (and POLES)
  - Next Lecture: Chapter 8

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### LECTURE OBJECTIVES

- ZEROS and POLES
- Relate  $H(z)$  to FREQUENCY RESPONSE

$$H(e^{j\hat{\omega}}) = H(z) \Big|_{z=e^{j\hat{\omega}}}$$

- THREE DOMAINS:
  - Show Relationship for FIR:

$$h[n] \leftrightarrow H(z) \leftrightarrow H(e^{j\hat{\omega}})$$

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### DESIGN PROBLEM

- Example:
  - Design a Lowpass FIR filter (Find  $b_k$ )
  - Reject completely  $0.7\pi$ ,  $0.8\pi$ , and  $0.9\pi$ 
    - This is NULLING
  - Estimate the filter length needed to accomplish this task. How many  $b_k$  ?
- Z POLYNOMIALS provide the TOOLS

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## Z-Transform DEFINITION

- POLYNOMIAL Representation of LTI SYSTEM:

$$H(z) = \sum_n h[n]z^{-n}$$

- EXAMPLE:

$$\{h[n]\} = \{2, 0, -3, 0, 2\}$$

$$\begin{aligned} H(z) &= 2z^{-0} + 0z^{-1} - 3z^{-2} + 0z^{-3} + 2z^{-4} \\ &= 2 - 3z^{-2} + 2z^{-4} \\ &= 2 - 3(z^{-1})^2 + 2(z^{-1})^4 \end{aligned}$$

APPLIES to Any SIGNAL

POLYNOMIAL in  $z^{-1}$

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## CONVOLUTION PROPERTY

- Convolution in the  $n$ -domain
  - SAME AS
- Multiplication in the  $z$ -domain

$$y[n] = h[n] * x[n] \leftrightarrow Y(z) = H(z)X(z)$$

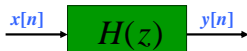
$$\begin{aligned} y[n] &= x[n] * h[n] \\ &= \sum_{k=0}^M h[k]x[n-k] \end{aligned}$$

FIR Filter

MULTIPLY z-TRANSFORMS

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## CONVOLUTION EXAMPLE



$$x[n] = \delta[n-1] + 2\delta[n-2] \quad h[n] = \delta[n] - \delta[n-1]$$

$$y[n] = x[n] * h[n]$$

$$X(z) = z^{-1} + 2z^{-2}$$

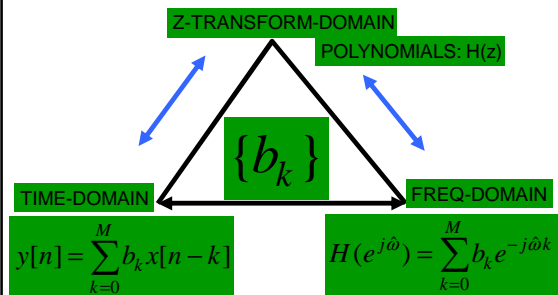
$$H(z) = 1 - z^{-1}$$

$$Y(z) = (z^{-1} + 2z^{-2})(1 - z^{-1}) = z^{-1} + z^{-2} - 2z^{-3}$$

$$y[n] = \delta[n-1] + \delta[n-2] - 2\delta[n-3]$$

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## THREE DOMAINS



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## FREQUENCY RESPONSE ?

- Same Form:

$\hat{\omega}$  - Domain

$$H(e^{j\hat{\omega}}) = \sum_{k=0}^M b_k e^{-j\hat{\omega}k}$$

$$H(e^{j\hat{\omega}}) = \sum_{k=0}^M b_k (e^{j\hat{\omega}})^{-k}$$

$$z = e^{j\hat{\omega}}$$

$z$  - Domain

$$H(z) = \sum_{k=0}^M b_k z^{-k}$$

SAME COEFFICIENTS

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## ANOTHER ANALYSIS TOOL

- $z$ -Transform POLYNOMIALS are EASY !
  - ROOTS, FACTORS, etc.
- ZEROS and POLES: where is  $H(z) = 0$  ?**
- The  $z$ -domain is COMPLEX
  - $H(z)$  is a COMPLEX-VALUED function of a COMPLEX VARIABLE  $z$ .

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### ZEROS of $H(z)$

- Find  $z$ , where  $H(z)=0$

$$H(z) = 1 - \frac{1}{2}z^{-1}$$

$$1 - \frac{1}{2}z^{-1} = 0 ?$$

$$z - \frac{1}{2} = 0$$

$$\text{Zero at : } z = \frac{1}{2}$$

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### ZEROS of $H(z)$

- Find  $z$ , where  $H(z)=0$

– Interesting when  $z$  is ON the unit circle.

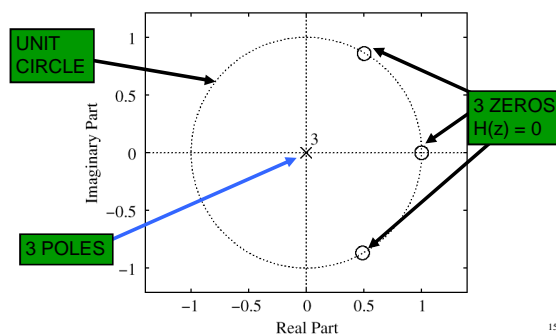
$$H(z) = 1 - 2z^{-1} + 2z^{-2} - z^{-3}$$

$$H(z) = (1 - z^{-1})(1 - z^{-1} + z^{-2})$$

$$\text{Roots : } z = 1, \frac{1}{2} \pm j\frac{\sqrt{3}}{2} \quad e^{\pm j\pi/3}$$

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### PLOT ZEROS in z-DOMAIN



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### POLES of $H(z)$

- Find  $z$ , where  $H(z) \rightarrow \infty$

– Not very interesting for the FIR case

$$H(z) = 1 - 2z^{-1} + 2z^{-2} - z^{-3}$$

$$H(z) = \frac{z^3 - 2z^2 + 2z - 1}{z^3}$$

$$\text{Three Poles at : } z = 0$$

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### FREQ. RESPONSE from ZEROS

$$H(e^{j\hat{\omega}}) = H(z) \Big|_{z=e^{j\hat{\omega}}}$$

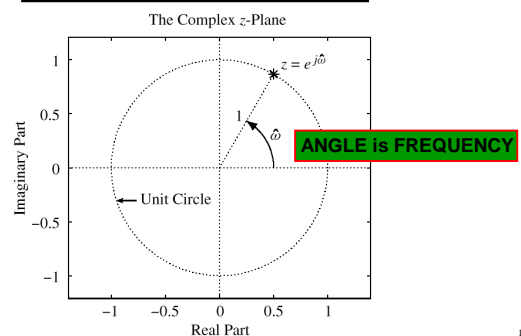
- Relate  $H(z)$  to FREQUENCY RESPONSE
  - EVALUATE  $H(z)$  on the **UNIT CIRCLE**
- ANGLE is same as FREQUENCY

$$z = e^{j\hat{\omega}} \quad (\text{as } \hat{\omega} \text{ varies})$$

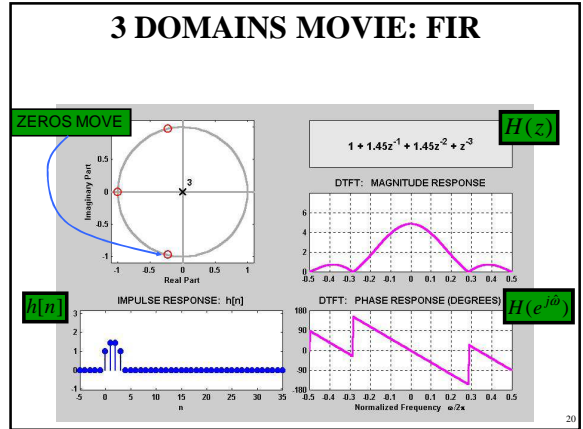
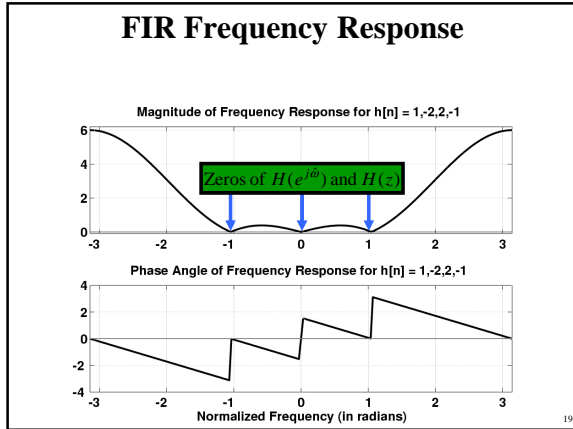
defines a **CIRCLE**, radius = 1

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$$H(e^{j\hat{\omega}}) = H(z) \Big|_{z=e^{j\hat{\omega}}}$$



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### NULLING PROPERTY of $H(z)$

- When  $H(z)=0$  on the unit circle.
  - Find inputs  $x[n]$  that give zero output

$$H(z) = 1 - 2z^{-1} + 2z^{-2} - z^{-3}$$

$$H(e^{j\hat{\omega}}) = 1 - 2e^{-j\hat{\omega}} + 2e^{-j2\hat{\omega}} - e^{-j3\hat{\omega}}$$

$x[n] \rightarrow$

$H(z)$

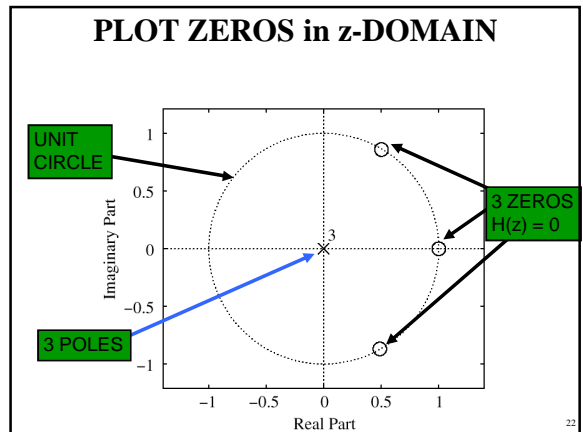
$\rightarrow y[n]$

$H(e^{j\pi/3}) = ?$

$x[n] = e^{j(\pi/3)n}$

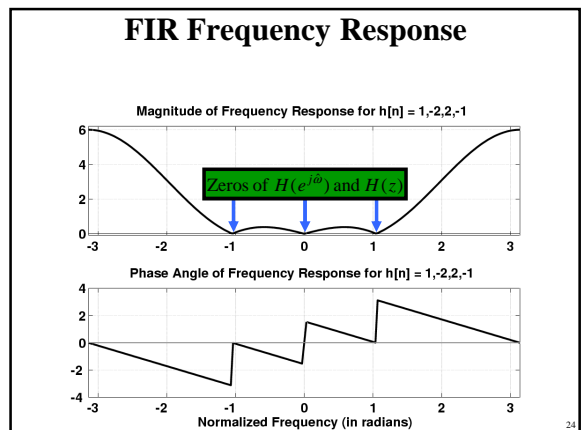
$\rightarrow$

$y[n] = H(e^{j(\pi/3)}) \cdot e^{j(\pi/3)n}$



### NULLING PROPERTY of $H(z)$

- Evaluate  $H(z)$  at the input "frequency"
  - $H(e^{j\hat{\omega}}) = 1 - 2e^{-j\hat{\omega}} + 2e^{-j2\hat{\omega}} - e^{-j3\hat{\omega}}$
  - $y[n] = H(e^{j\pi/3}) \cdot e^{j(\pi/3)n}$
  - $y[n] = (1 - 2e^{-j\pi/3} + 2e^{-j2\pi/3} - e^{-j3\pi/3}) \cdot e^{j(\pi/3)n}$
  - $(1 - 2(\frac{1}{2} - j\frac{\sqrt{3}}{2}) + 2(-\frac{1}{2} - j\frac{\sqrt{3}}{2}) - (-1))$
  - $y[n] = (1 - 1 + j\sqrt{3} - 1 - j\sqrt{3} + 1) \cdot e^{j(\pi/3)n} = 0$

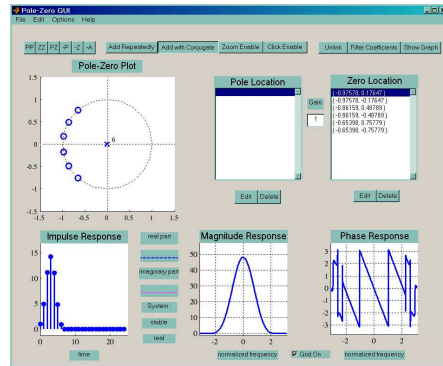


## DESIGN PROBLEM

- Example:
  - Design a Lowpass FIR filter (Find  $b_k$ )
  - Reject completely  $0.7\pi$ ,  $0.8\pi$ , and  $0.9\pi$
  - Estimate the filter length needed to accomplish this task. How many  $b_k$ ?
- Z POLYNOMIALS provide the TOOLS

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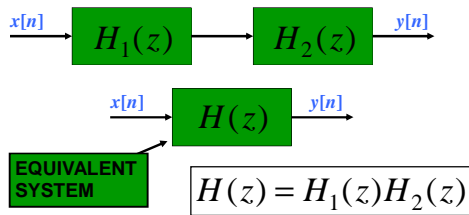
## PeZ Demo: Zero Placing



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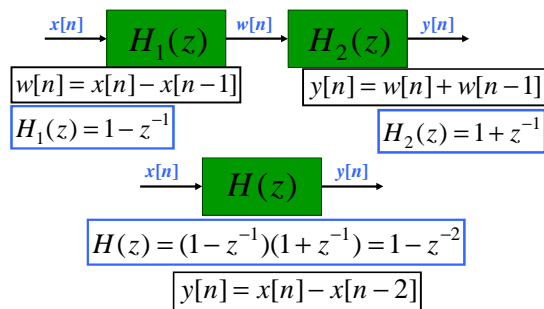
## CASCADE EQUIVALENT

- Multiply the System Functions



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## CASCADE EXAMPLE



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## L-pt RUNNING SUM $H(z)$

$$H(z) = \sum_{k=0}^{L-1} \frac{1}{L} z^{-k} = \frac{1 - z^{-L}}{L(1 - z^{-1})} = \frac{z^L - 1}{Lz^{L-1}(z - 1)}$$

$$z^L - 1 = 0 \Rightarrow z^L = 1 = e^{j2\pi k}$$

$$z = e^{j(2\pi/L)k} \text{ for } k = 1, 2, \dots, L-1$$

ZEROS on UNIT CIRCLE

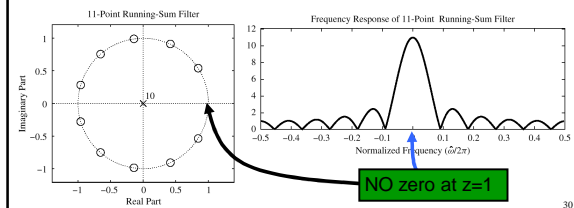
$(z-1)$  in denominator cancels  $k=0$  term

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## 11-pt RUNNING SUM $H(z)$

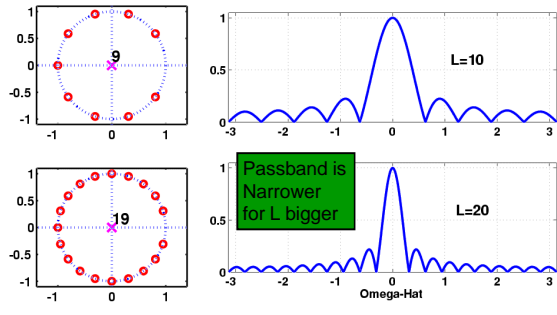
$$H(z) = \sum_{k=0}^{10} z^{-k}$$

$$H(z) = (1 - e^{j2\pi/11}z^{-1})(1 - e^{j4\pi/11}z^{-1}) \dots (1 - e^{j20\pi/11}z^{-1})$$



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### FILTER DESIGN: CHANGE L



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### NULLING FILTER

- PLACE ZEROS to make  $y[n] = 0$

$$H(z) = 1 - 2z^{-1} + 2z^{-2} - z^{-3}$$

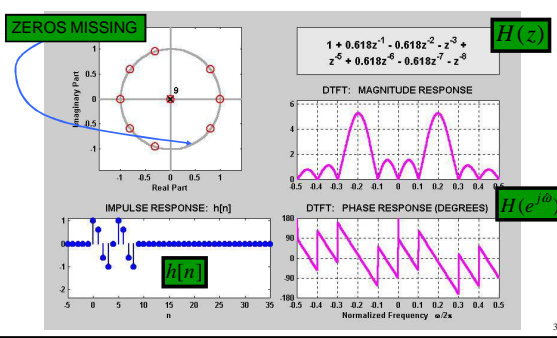
3 ZEROS  
 $H(z) = 0$

the output resulting from each of the following three signals will be zero:

$H(z_1) = 0$	$x_1[n] = (z_1)^n = 1$	$\rightarrow$	$y_1[n] = 0$
$H(z_2) = 0$	$x_2[n] = (z_2)^n = e^{j\pi n/3}$		$y_2[n] = 0$
$H(z_3) = 0$	$x_3[n] = (z_3)^n = e^{-j\pi n/3}$		$y_3[n] = 0$

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### 3 DOMAINS MOVIE: FIR



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### POP QUIZ: MAG & PHASE

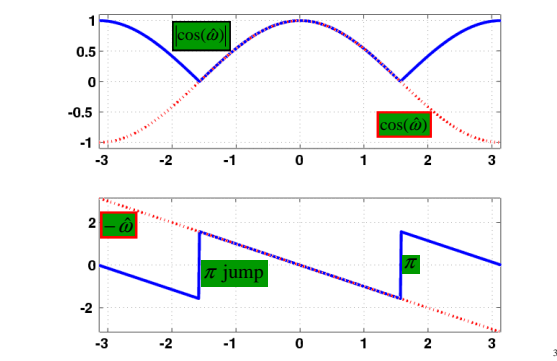
- Given:  $H(e^{j\hat{\omega}}) = e^{-j\hat{\omega}} \cos(\hat{\omega})$
- Derive Magnitude and Phase

$$|H(e^{j\hat{\omega}})| = |e^{-j\hat{\omega}}| \cdot |\cos(\hat{\omega})| = |\cos(\hat{\omega})|$$

$$\angle H(e^{j\hat{\omega}}) = \begin{cases} -\hat{\omega} & \cos(\hat{\omega}) \geq 0 \\ -\hat{\omega} + \pi & \cos(\hat{\omega}) < 0 \end{cases}$$

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### Ans: FREQ RESPONSE



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### POP QUIZ : Answer #2

- Find  $y[n]$  when

$$x[n] = \cos(0.25\pi n)$$

$$y[n] = |H| \cos(0.25\pi n + \angle H)$$

$$= 0.707 \cos(0.25\pi n - \frac{\pi}{4})$$

$$H(e^{j\hat{\omega}}) = e^{-j\hat{\omega}} \cos(\hat{\omega}) \quad \text{at } \hat{\omega} = \frac{\pi}{4}$$

$$H(e^{j\pi/4}) = e^{-j\pi/4} \cos(\frac{\pi}{4}) = 0.707 e^{-j\pi/4}$$

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## CHANGE in NOTATION

- Relate  $H(z)$  to FREQUENCY RESPONSE

$$H(\hat{\omega}) = H(z) \Big|_{z=e^{j\hat{\omega}}}$$

NEW NOTATION for FREQUENCY RESPONSE

$$H(\hat{\omega}) \leftrightarrow H(e^{j\hat{\omega}})$$

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