

Digital Signal Processing

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Digital Signal Processing

Lecture 12

Frequency Response of FIR Filters

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READING ASSIGNMENTS

- This Lecture:
 - Chapter 6, Sections 6-1, 6-2, 6-3, 6-4, & 6-5
- Other Reading:
 - Recitation: Chapter 6
 - FREQUENCY RESPONSE EXAMPLES
 - Next Lecture: Chap. 6, Sects. 6-6, 6-7 & 6-8

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LECTURE OBJECTIVES

- **SINUSOIDAL INPUT SIGNAL**
 - DETERMINE the FIR FILTER OUTPUT
- **FREQUENCY RESPONSE** of FIR
 - PLOTTING vs. Frequency
 - MAGNITUDE vs. Freq
 - PHASE vs. Freq

$$H(e^{j\hat{\omega}}) = \left| H(e^{j\hat{\omega}}) \right| e^{j\angle H(e^{j\hat{\omega}})}$$

MAG

PHASE

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DOMAINS: Time & Frequency

- **Time-Domain: “n” = time**
 - $x[n]$ discrete-time signal
 - $x(t)$ continuous-time signal
- **Frequency Domain (sum of sinusoids)**
 - Spectrum vs. f (Hz)
 - ANALOG vs. DIGITAL
 - Spectrum vs. omega-hat
- Move back and forth **QUICKLY**

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LTI SYSTEMS

- LTI: **L**inear & **T**ime-**I**nvariant
- COMPLETELY CHARACTERIZED by:
 - **IMPULSE RESPONSE** $h[n]$
 - **CONVOLUTION**: $y[n] = x[n] * h[n]$
 - The "rule" defining the system can ALWAYS be re-written as convolution
- FIR Example: $h[n]$ is same as b_k

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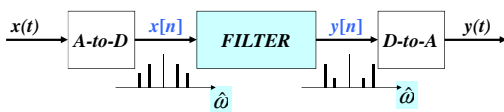
POP QUIZ

- FIR Filter is "FIRST DIFFERENCE"
 - $y[n] = x[n] - x[n-1]$
- Write output as a convolution
 - Need impulse response
 - $h[n] = \delta[n] - \delta[n-1]$
 - Then, another way to compute the output:

$$y[n] = (\delta[n] - \delta[n-1]) * x[n]$$

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DIGITAL "FILTERING"



- CONCENTRATE on the **SPECTRUM**
- SINUSOIDAL INPUT
 - INPUT $x[n] = \text{SUM}$ of SINUSOIDS
 - Then, OUTPUT $y[n] = \text{SUM}$ of SINUSOIDS

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FILTERING EXAMPLE

- 7-point AVERAGER
 - Removes cosine
 - By making its amplitude (A) smaller
- 3-point AVERAGER
 - Changes A slightly

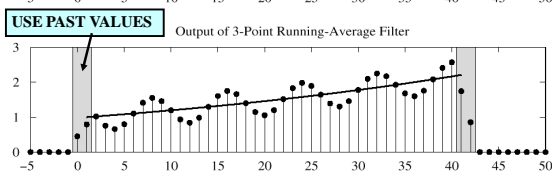
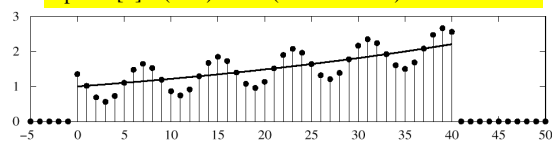
$$y_7[n] = \sum_{k=0}^6 \left(\frac{1}{7}\right) x[n-k]$$

$$y_3[n] = \sum_{k=0}^2 \left(\frac{1}{3}\right) x[n-k]$$

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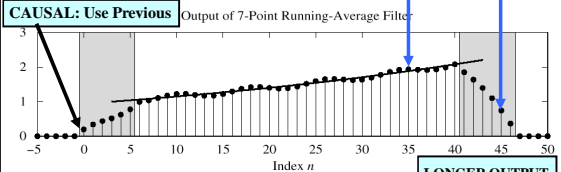
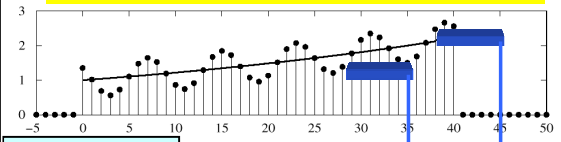
3-pt AVG EXAMPLE

Input : $x[n] = (1.02)^n + \cos(2\pi n/8 + \pi/4)$ for $0 \leq n \leq 40$



7-pt FIR EXAMPLE (AVG)

Input : $x[n] = (1.02)^n + \cos(2\pi n/8 + \pi/4)$ for $0 \leq n \leq 40$

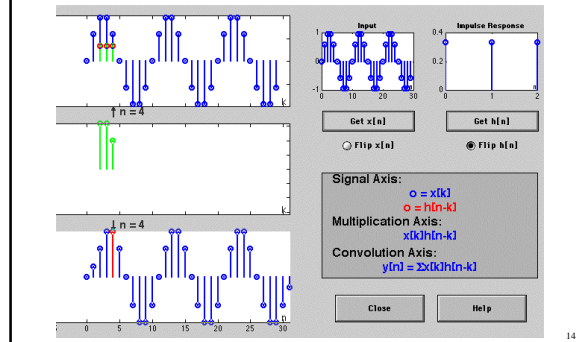


SINUSOIDAL RESPONSE

- INPUT: $x[n] = \text{SINUSOID}$
- OUTPUT: $y[n]$ will also be a SINUSOID
 - Different Amplitude and Phase
 - **SAME** Frequency
- AMPLITUDE & PHASE CHANGE
 - Called the **FREQUENCY RESPONSE**

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DCONVDEMO: MATLAB GUI



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COMPLEX EXPONENTIAL

$$x[n] = Ae^{j\varphi} e^{j\hat{\omega}n} \quad -\infty < n < \infty$$

$x[n]$ is the input signal—a complex exponential

$$y[n] = \sum_{k=0}^M b_k x[n-k] = \sum_{k=0}^M h[k] x[n-k]$$

FIR DIFFERENCE EQUATION

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COMPLEX EXP OUTPUT

- Use the FIR “Difference Equation”

$$\begin{aligned} y[n] &= \sum_{k=0}^M b_k x[n-k] = \sum_{k=0}^M b_k Ae^{j\varphi} e^{j\hat{\omega}(n-k)} \\ &= \left(\sum_{k=0}^M b_k e^{j\hat{\omega}(-k)} \right) Ae^{j\varphi} e^{j\hat{\omega}n} \\ &= H(\hat{\omega}) Ae^{j\varphi} e^{j\hat{\omega}n} \end{aligned}$$

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FREQUENCY RESPONSE

- At each frequency, we can **DEFINE**

$$H(e^{j\hat{\omega}}) = \sum_{k=0}^M b_k e^{-j\hat{\omega}k}$$

FREQUENCY RESPONSE

- Complex-valued formula
 - Has **MAGNITUDE** vs. frequency
 - And **PHASE** vs. frequency
- Notation: $H(e^{j\hat{\omega}})$ in place of $H(\hat{\omega})$

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EXAMPLE 6.1

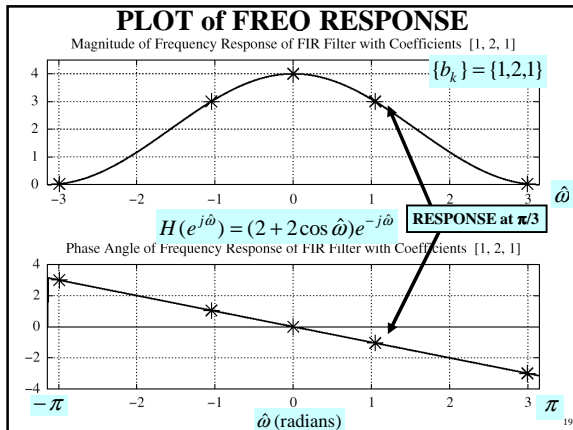
$$\{b_k\} = \{1, 2, 1\}$$

$$\begin{aligned} H(e^{j\hat{\omega}}) &= 1 + 2e^{-j\hat{\omega}} + e^{-j2\hat{\omega}} \\ &= e^{-j\hat{\omega}}(e^{j\hat{\omega}} + 2 + e^{-j\hat{\omega}}) \\ &= e^{-j\hat{\omega}}(2 + 2\cos \hat{\omega}) \end{aligned}$$

EXPLOIT SYMMETRY

Since $(2 + 2\cos \hat{\omega}) \geq 0$
 Magnitude is $|H(e^{j\hat{\omega}})| = (2 + 2\cos \hat{\omega})$
 and Phase is $\angle H(e^{j\hat{\omega}}) = -\hat{\omega}$

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EXAMPLE 6.2

Find $y[n]$ when $H(e^{j\hat{\omega}})$ is known and $x[n] = 2e^{j\pi/4}e^{j(\pi/3)n}$

$H(e^{j\hat{\omega}}) = (2 + 2 \cos \hat{\omega})e^{-j\hat{\omega}}$

EXAMPLE 6.2 (answer)

Find $y[n]$ when $x[n] = 2e^{j\pi/4}e^{j(\pi/3)n}$

One Step - evaluate $H(e^{j\hat{\omega}})$ at $\hat{\omega} = \pi/3$

$H(e^{j\hat{\omega}}) = (2 + 2 \cos \hat{\omega})e^{-j\hat{\omega}}$

$H(e^{j\hat{\omega}}) = 3e^{-j\pi/3} \quad @ \hat{\omega} = \pi/3$

$y[n] = (3e^{-j\pi/3}) \times 2e^{j\pi/4}e^{j(\pi/3)n} = 6e^{-j\pi/12}e^{j(\pi/3)n}$

EXAMPLE: COSINE INPUT

Find $y[n]$ when $H(e^{j\hat{\omega}})$ is known and $x[n] = 2 \cos(\frac{\pi}{3}n + \frac{\pi}{4})$

$H(e^{j\hat{\omega}}) = (2 + 2 \cos \hat{\omega})e^{-j\hat{\omega}}$

EX: COSINE INPUT

Find $y[n]$ when $x[n] = 2 \cos(\frac{\pi}{3}n + \frac{\pi}{4})$

$2 \cos(\frac{\pi}{3}n + \frac{\pi}{4}) = e^{j(\pi n/3 + \pi/4)} + e^{-j(\pi n/3 + \pi/4)}$

$\Rightarrow x[n] = x_1[n] + x_2[n]$

Use Linearity

$y_1[n] = H(e^{j\pi/3})e^{j(\pi n/3 + \pi/4)}$

$y_2[n] = H(e^{-j\pi/3})e^{-j(\pi n/3 + \pi/4)}$

$\Rightarrow y[n] = y_1[n] + y_2[n]$

EX: COSINE INPUT (ans-2)

Find $y[n]$ when $x[n] = 2 \cos(\frac{\pi}{3}n + \frac{\pi}{4})$

$H(e^{j\hat{\omega}}) = (2 + 2 \cos \hat{\omega})e^{-j\hat{\omega}}$

$y_1[n] = H(e^{j\pi/3})e^{j(\pi n/3 + \pi/4)} = 3e^{-j(\pi/3)}e^{j(\pi n/3 + \pi/4)}$

$y_2[n] = H(e^{-j\pi/3})e^{-j(\pi n/3 + \pi/4)} = 3e^{j(\pi/3)}e^{-j(\pi n/3 + \pi/4)}$

$y[n] = 3e^{j(\pi n/3 - \pi/12)} + 3e^{-j(\pi n/3 - \pi/12)}$

$\Rightarrow y[n] = 6 \cos(\frac{\pi}{3}n - \frac{\pi}{12})$

MATLAB: FREQUENCY RESPONSE

- **HH = freqz(bb,1,ww)**
 - VECTOR **bb** contains Filter Coefficients
 - SP-First: **HH = freqz(bb,1,ww)**
- FILTER COEFFICIENTS $\{b_k\}$

$$H(e^{j\hat{\omega}}) = \sum_{k=0}^M b_k e^{-j\hat{\omega}k}$$

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LTI SYSTEMS

- LTI: Linear & Time-Invariant
- COMPLETELY CHARACTERIZED by:
 - FREQUENCY RESPONSE, or
 - IMPULSE RESPONSE $h[n]$
- Sinusoid IN -----> Sinusoid OUT
 - At the SAME Frequency

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Time & Frequency Relation

- Get Frequency Response from $h[n]$
 - Here is the FIR case:

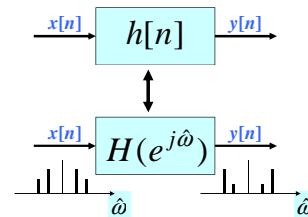
$$H(e^{j\hat{\omega}}) = \sum_{k=0}^M b_k e^{-j\hat{\omega}k} = \sum_{k=0}^M h[k] e^{-j\hat{\omega}k}$$

IMPULSE RESPONSE

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BLOCK DIAGRAMS

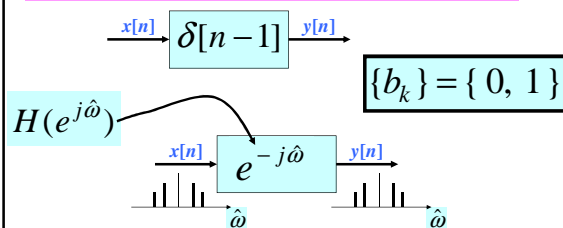
- Equivalent Representations



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UNIT-DELAY SYSTEM

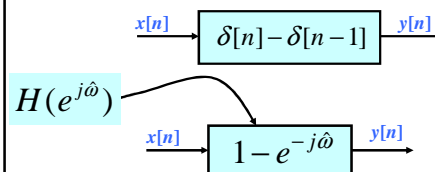
Find $h[n]$ and $H(e^{j\hat{\omega}})$ for $y[n] = x[n-1]$



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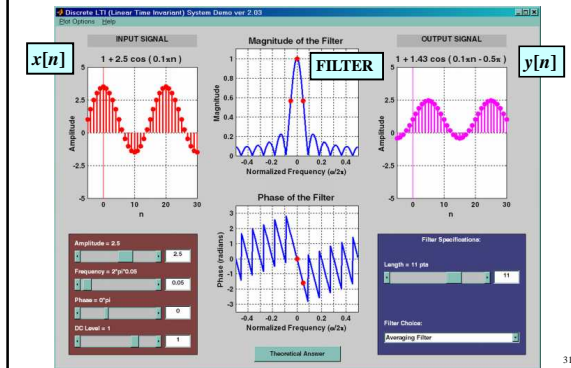
FIRST DIFFERENCE SYSTEM

Find $h[n]$ and $H(e^{j\hat{\omega}})$ for the Difference Equation: $y[n] = x[n] - x[n-1]$



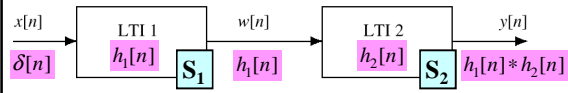
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DLTI Demo with Sinusoids



CASCADE SYSTEMS

- Does the order of S_1 & S_2 matter?
 - NO, LTI SYSTEMS can be rearranged !!!
 - WHAT ARE THE FILTER COEFFS? $\{b_k\}$
 - WHAT is the overall FREQUENCY RESPONSE ?



CASCADE EQUIVALENT

- MULTIPLY the Frequency Responses

